

Lecture 4: Small-scale fading: Statistical Multipath Channel Models

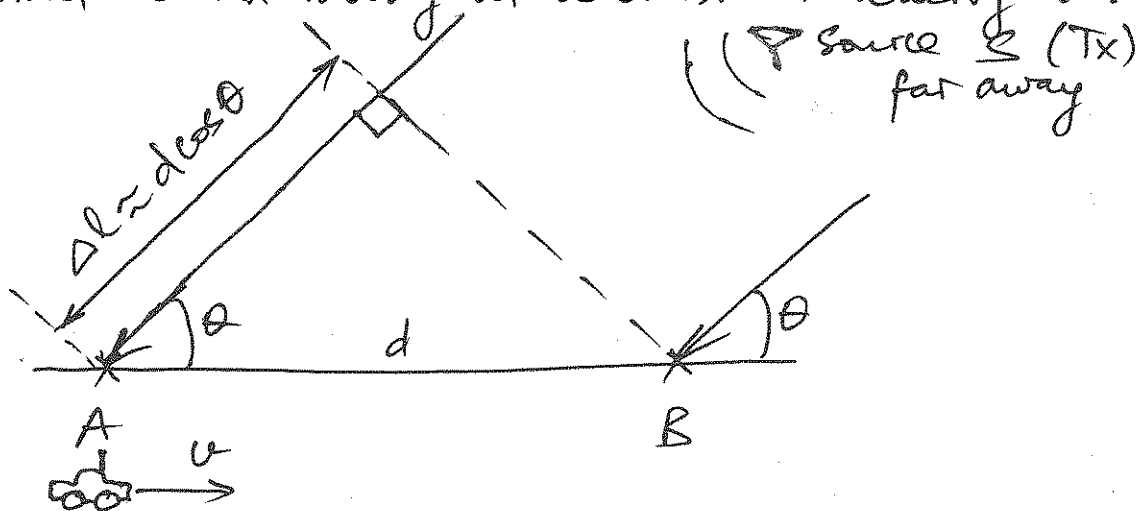
- As discussed earlier, the actual received signal/power at a receive antenna varies significantly over time, even at a given Tx-Rx distance and a given frequency band.
 - This fluctuation in the received signal power is caused by reflection and scattering effects.
 - Usually the number of scatterers is large, causing the signal to arrive at the receive antenna via multiple propagation paths.
 - These multipaths can add constructively and destructively, causing the signal at the receiver to appear random.
 - This randomness in received signal is called small-scale fading (caused mainly by scatterers in the vicinity of the receiver) and is usually characterized statistically.
- Small-scale fading effects include:
- rapid changes in signal strength over a small distance or time interval.
 - random frequency expansion due to the Doppler effect
 - time dispersion (echos) caused by multipath propagation delays.

The received signal at any point in space may consist of a large number of planewaves (paths) with random amplitudes, phases, and angles of arrival.

- +) Factors influencing small-scale fading:
- o multipath propagation from reflection and scattering objects
 - o speed of the Tx and/or Rx
 - o speed of surrounding objects
 - o the transmission bandwidth of the signal.

4) Doppler shift:

Consider a Rx moving at a constant velocity v .



Assume source S (Tx) is far away such that far field propagation holds between point A & B.

Suppose:

- d = distance traveled between A & B.
- θ = angle of arrival
- v = speed of moving object (Rx)
- Δt = time required for traveling from A \rightarrow B.

Then there will be a change in the phase of the received signal due to mobility. The phase change is proportional to the travel distance measured in wavelengths:

$$\Delta \phi = \frac{2\pi \Delta l}{\lambda} = \frac{2\pi d \cos \theta}{\lambda} = \frac{2\pi}{\lambda} v \cdot \Delta t \cdot \cos \theta$$

This phase change will cause a change in the instantaneous frequency of the received signal. Specifically the frequency is shifted by an amount called the "Doppler" shift, given as

$$f_d = \frac{1}{2\pi} \frac{\Delta\phi}{\Delta t} = \frac{v}{\lambda} \cos\theta, \quad -\pi \leq \theta \leq \pi.$$

• If $|\theta| < \frac{\pi}{2}$: Rx moving towards Tx $\rightarrow \cos\theta > 0$
 $f_d > 0$ or the frequency is increased.

$|\theta| > \frac{\pi}{2}$: Rx moving away from Tx $\rightarrow \cos\theta < 0$
 $f_d < 0$ or the frequency is decreased.

This is the effect you hear on a siren, the pitch increases as the siren moves towards you and then decreases as it moves away from you.

• For moving Tx/Rx, the Doppler effect creates an expansion in the frequency of the received signal that is called the Doppler spread.

If the transmitted signal were a pure sinusoidal signal (single tone), it will be received spread over a range of frequencies.

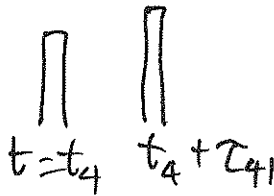
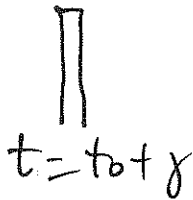
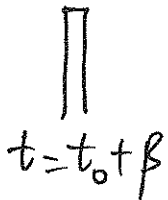
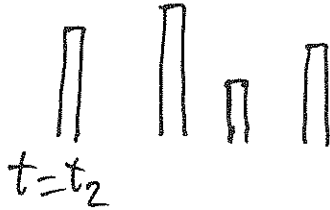
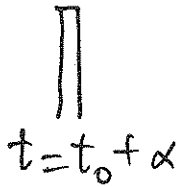
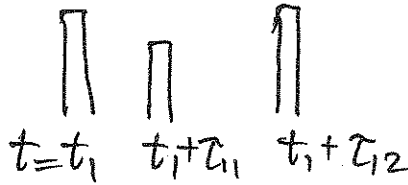
2.1. Characterization of fading multipath channels:

• We will now characterize the channel by its canonical form: the impulse response.

• Suppose that we transmit extremely short pulses, ideally impulses, over a multipath fading channel.

Tx signal

Rx signal



- The received signal may appear as a train of pulses with:
 - varying number of pulses
 - varying strength of individual pulses
 - varying spacing among the pulses

These variations appear unpredictable. Thus multipath time-varying channel is often characterized statistically.

- Suppose that we transmit a signal that is represented in general as

$$s(t) = \text{Re} \left\{ s_e(t) e^{j2\pi f_c t} \right\} \quad (1)$$

where f_c : carrier frequency

$s_e(t)$: baseband (lowpass) equivalent signal
(the part that actually contains information)

Assume multipath propagation, each path has a certain propagation delay and an attenuation factor.

◦ The received signal can then be expressed as

$$r(t) = \sum_n \alpha_n(t) s[t - \tau_n(t)] \quad (2)$$

where:

$$\begin{aligned} n &= \text{path number (index)} \\ \alpha_n(t) &= \text{attenuation factor for the } n^{\text{th}} \text{ path} \\ \tau_n(t) &= \text{propagation delay} \end{aligned}$$

◦ Putting (1) and (2) together yields

$$r(t) = \text{Re} \left\{ \underbrace{\left(\sum_n \alpha_n(t) s_e[t - \tau_n(t)] \right)}_{\text{equivalent low-pass received signal (baseband)}} e^{-j2\pi f_c \tau_n(t)} e^{j2\pi f_c t} \right\}$$

◦ Thus the equivalent baseband channel has the following impulse response:

$$h(\tau; t) = \sum_n \alpha_n(t) e^{-j2\pi f_c \tau_n(t)} \cdot \delta(t - \tau_n(t)) \quad (3)$$

◦ For channels with many scatterers and multipath, we can also express the received signal on a continuum of multipath components. We re-write (2) as

$$r(t) = \int \alpha(\tau; t) s(t - \tau) d\tau \quad (4)$$

where

$\alpha(\tau; t)$: attenuation of the signal components at delay τ .

Then the equivalent baseband impulse response becomes

$$h(\tau; t) = \alpha(\tau; t) e^{-j2\pi f_c \tau} \quad (5)$$

and we can express the received signal via standard convolution:

$$r(t) = \text{Re} \left\{ \left[\underbrace{h(\tau; t) * s_e(t)}_{\text{baseband equivalent received signal}} \right] \cdot e^{j2\pi f_c t} \right\}$$

where

$$h(\tau; t) * s_e(t) = \int_{-\infty}^{\infty} h(\tau; t) s_e(t - \tau) d\tau.$$

o The impulse response of a fading channel $h(\tau; t)$ as given in (3) and (5) is linear and time-varying; (in contrast to LTI systems which are time-invariant).

To understand the effect of a linear time-varying channel, let's perform a simple thought experiment.

Suppose we send a constant signal, $s_e(t) = 1 \forall t$. Then the baseband equivalent received signal is

$$r_e(t) = \sum_n \alpha_n(t) e^{-j2\pi f_c \tau_n(t)} = \sum_n \alpha_n(t) e^{j\theta_n(t)}$$

where $\alpha_n(t)$: path amplitude (of the n^{th} path)
 $\theta_n(t)$: path phase

Both path amplitude and phase depend on the dynamics in the channel (or propagation environment/medium).

• Changes in $\alpha_n(t)$ usually requires large dynamic changes in the propagation medium.

• Changes in $\theta_n(t)$ however can be very rapid: $\theta_n(t)$ will change by 2π whenever the path delay τ_n changes by $\frac{1}{f_c}$, which is a very small number esp. at large f_c carrier frequencies.

$\theta_n(t)$ will change by 2π rad with relatively small motion in the medium.

Path delays $\tau_n(t)$ also changes in a unpredictable (random) manner.

• Thus the channel impulse response $h(\tau; t)$ can be viewed as a complex-valued random process.

+ Signal fading: Multipath propagation results in signal fading, which is primarily the results of the time variation in the phases $\theta_n(t)$.

Sometimes these different paths add destructively, leading to the very weak received signals.

Sometimes they can add constructively, leading to a strong received signal.

This amplitude variation in the received signal is termed "signal fading".

• We will study the statistical models of fading channels but first we will examine the properties of a linear time-varying channel, without paying attention to the fact that it is random.